

Lecture 10
EE 2303/001-Electronics I
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Professor Ronald L. Carter
ronc@uta.edu
<http://www.uta.edu/ronc/>

SPICE Diode Static Model Eqns

$$I_d = area * (I_{fwd} - I_{rev})$$

$$I_{fwd} = I_{nrm} * K_{inj} + I_{rec} * K_{gen}$$

$$I_{nrm} = I_S * \{ \exp [V_d / (N * V_t)] - 1 \}$$

K_{inj} = high-injection factor

For $IKF > 0$, $K_{inj} = IKF / [IKF + I_{nrm}]^{1/2}$
otherwise, $K_{inj} = 1$

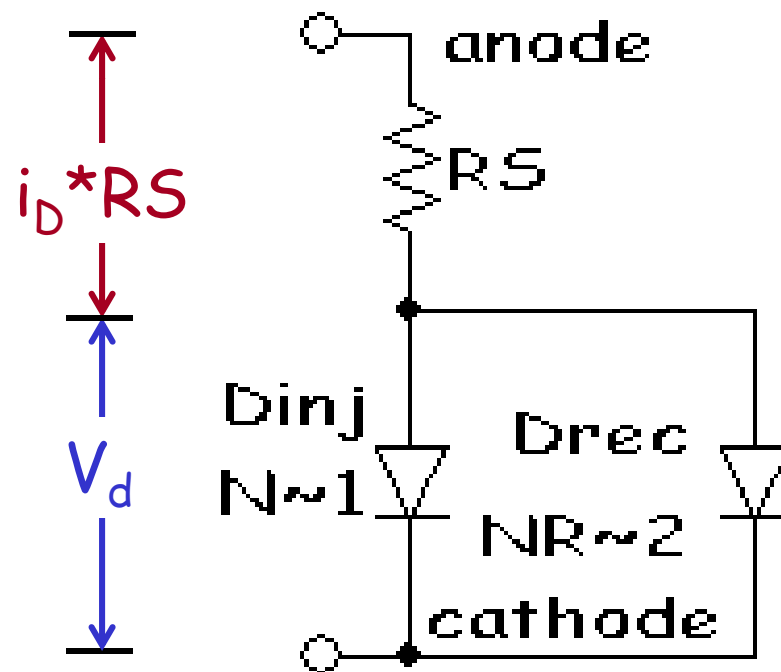
$$I_{rec} = I_{SR} * \{ \exp [V_d / (N_R * V_t)] - 1 \}$$

$$K_{gen} = ((1 - V_d / V_J)^2 + 0.005)^{M/2}$$

SPICE Diode Static Model

- D_{inj}
 - I_S
 - $N \sim 1$
 - $I_{KF}, V_{KF}, N \sim 1$
- D_{rec}
 - I_{SR}
 - $NR \sim 2$

$$V_{ext} = V_D + i_D * R_S$$



D Diode (See [1])

General Form

D<*name*> <(+) *node*> <(-) *node*> <*model name*> [*area value*]

Examples

```
DCLAMP 14 0 DMOD
D13 15 17 SWITCH 1.5
```

Model Form

```
.MODEL <model name> D [model parameters]
.model D1N4148-X D(Is=2.682n N=1.836
Rs=.5664 Ikf=44.17m Xti=3 Eg=1.11
Cj o=4p M=.3333 Vj=.5 Fc=.5
Isr=1.565n Nr=2 Bv=100 Ibv=10 Ou
Tt=11.54n)
*$
```

Diode Model Parameters (See [1])

Model Parameters (see .MODEL statement)

	Description	Unit	
	Default		
IS	Saturation current	amp	1E-14
N	Emission coefficient		1
ISR	Recombination current parameter	amp	0
NR	Emission coefficient for ISR		1
IKF	High-injection “knee” current	amp	infinite
BV	Reverse breakdown “knee” voltage	volt	infinite
IBV	Reverse breakdown “knee” current	amp	1E-10
NBV	Reverse breakdown ideality factor		1
RS	Parasitic resistance	ohm	0
TT	Transit time	sec	0
CJO	Zero-bias <i>p-n</i> capacitance	farad	0
VJ	<i>p-n</i> potential	volt	1
M	<i>p-n</i> grading coefficient		0.5
FC	Forward-bias depletion cap. coef,		0.5
EG	Bandgap voltage (barrier height)	eV	1.11

Diode Model Parameters (See [1])

Model Parameters (see .MODEL statement)

	Description	Unit	
	Default		
XTI	IS temperature exponent		3
TIKF	IKF temperature coefficient (linear)	°C ⁻¹	0
TBV1	BV temperature coefficient (linear)	°C ⁻¹	0
TBV2	BV temperature coefficient (quadratic)	°C ⁻²	0
TRS1	RS temperature coefficient (linear)	°C ⁻¹	0
TRS2	RS temperature coefficient (quadratic)	°C ⁻²	0
T_MEASURED	Measured temperature	°C	
T_ABS	Absolute temperature	°C	
T_REL_GLOBAL	Rel. to curr. Temp.	°C	
T_REL_LOCAL	Relative to AKO model temperature	°C	

For information on **T_MEASURED**, **T_ABS**, **T_REL_GLOBAL**, and **T_REL_LOCAL**, see the .MODEL statement.

The diode is modeled as an ohmic resistance (**RS/area**) in series with an intrinsic diode. $\langle(+)$ *node* \rangle is the anode and $\langle(-)$ *node* \rangle is the cathode. Positive current is current flowing from the anode through the diode to the cathode. [*area value*] scales **IS**, **ISR**, **IKF**, **RS**, **CJO**, and **IBV**, and defaults to 1. **IBV** and **BV** are both specified as positive values. In the following equations:

V_d = voltage across the intrinsic diode only

V_t = $k \cdot T / q$ (thermal voltage)

k = Boltzmann's constant

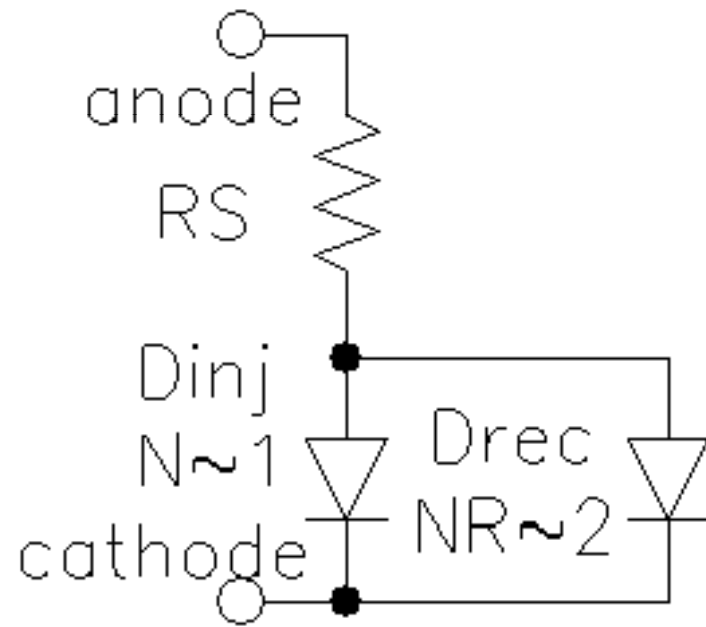
q = electron charge

T = analysis temperature ($^{\circ}\text{K}$)

T_{nom} = nom. temp. (set with TNOM option)

SPIICE Diode Model

- D_{inj}
 - $N \sim 1, r_d \sim N * V_t / i_D$
 - $r_d * C_d = TT = \tau$
 - C_{depl} given by C_{JO}, V_J and M
- D_{rec}
 - $N \sim 2, r_d \sim N * V_t / i_D$
 - $r_d * C_d = ?$
 - $C_{depl} = ?$



DC Current (See [1])

$$I_d = area * (I_{fwd} - I_{rev})$$

$$I_{fwd} = \text{forward current} = I_{nrm} * K_{inj} + I_{rec} * K_{gen}$$

$$I_{nrm} = \text{normal current} = I_S * (\exp(V_d / (N * V_t)) - 1)$$

K_{inj} = high-injection factor

$$\text{For: } IKF > 0, K_{inj} = (IKF / (IKF + I_{nrm}))^{1/2}$$

otherwise, $K_{inj} = 1$

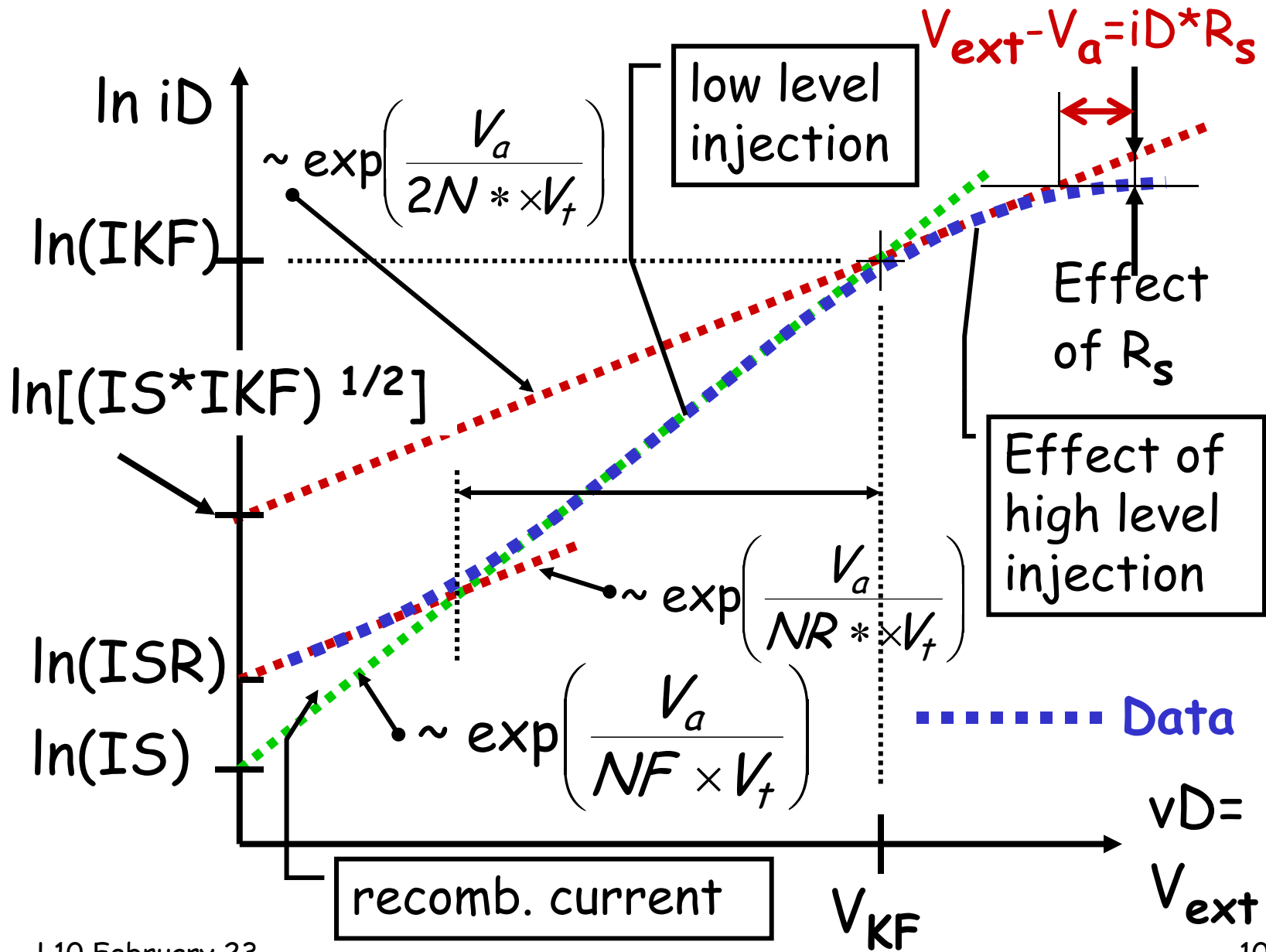
$$I_{rec} = \text{rec. cur.} = I_{SR} * (\exp(V_d / (NR * V_t)) - 1)$$

$$K_{gen} = \text{generation factor} = ((1 - V_d / V_J)^2 + 0.005)^{M/2}$$

$$I_{rev} = \text{reverse current} = I_{rev_high} + I_{rev_low}$$

$$I_{rev_high} = IBV * \exp[-(V_d + BV) / (NBV * V_t)]$$

$$I_{rev_low} = IBVL * \exp[-(V_d + BV) / (NBVL * V_t)]$$



Interpreting a plot of $\log(i_D)$ vs. V_d

In the region where $I_{rec} < I_{nrm} < I_{KF}$,
and $i_D \cdot R_S \ll V_d$.

$$i_D \sim I_{nrm} = I_S \cdot (\exp(V_d / (N \cdot V_t)) - 1)$$

For $N = 1$ and $V_t = 25.852 \text{ mV}$, the slope of the plot of $\log(i_D)$ vs. V_d is evaluated as

$$\begin{aligned} \{d\log(i_D)/dV_d\} &= \log(e) / (N \cdot V_t) \\ &= 16.799 \text{ decades/V} \\ &= 1 \text{ decade} / 59.526 \text{ mV} \end{aligned}$$

Static Model Eqns. Parameter Extraction

In the region where $I_{rec} < I_{nrm} < IKF$,
and $i_D * R_S \ll V_d$.

$$i_D \sim I_{nrm} = I_S * (\exp(V_d / (N * V_t)) - 1)$$

$$\{di_D / dV_d\} / i_D = d[\ln(i_D)] / dV_d = 1 / (N * V_t)$$

so $N \sim \{dV_d / d[\ln(i_D)]\} / V_t \equiv N_{eff},$

and $\ln(I_S) \sim \ln(i_D) - V_d / (N * V_t) \equiv \ln(I_{S_{eff}}).$

Note: i_D , V_t , etc., are normalized to 1A, 1V, resp.

Static Model Eqns. Parameter Extraction

In the region where $I_{rec} > I_{nrm}$, and $i_D * R_S \ll V_d$.

$$i_D \sim I_{rec} = I_{SR} * (\exp(V_d / (N_R * V_t)) - 1)$$

$$\{di_D / dV_d\} / i_D = d[\ln(i_D)] / dV_d \sim 1 / (N_R * V_t)$$

so $N_R \sim \{dV_d / d[\ln(i_D)]\} / V_t \equiv N_{eff},$

& $\ln(I_{SR}) \sim \ln(i_D) - V_d / (N_R * V_t) \equiv \ln(I_{SR_{eff}}).$

Note: i_D , V_t , etc., are normalized to 1A, 1V, resp.

Static Model Eqns. Parameter Extraction

In the region where $IKF > I_{nrm}$, and $i_D * R_S \ll V_d$.

$$i_D \sim [IS * IKF]^{1/2} * (\exp(V_d / (2 * N * V_t)) - 1)$$

$$\{di_D / dV_d\} / i_D = d[\ln(i_D)] / dV_d \sim (2 * N * V_t)^{-1}$$

so $2N \sim \{dV_d / d[\ln(i_D)]\} / V_t \equiv 2N_{eff},$

$$\text{and } \ln(i_D) - V_d / (NR * V_t) \equiv \frac{1}{2} \ln(IS * IKF_{eff}).$$

Note: i_D , V_t , etc., are normalized to 1A, 1V, resp.

Static Model Eqns. Parameter Extraction

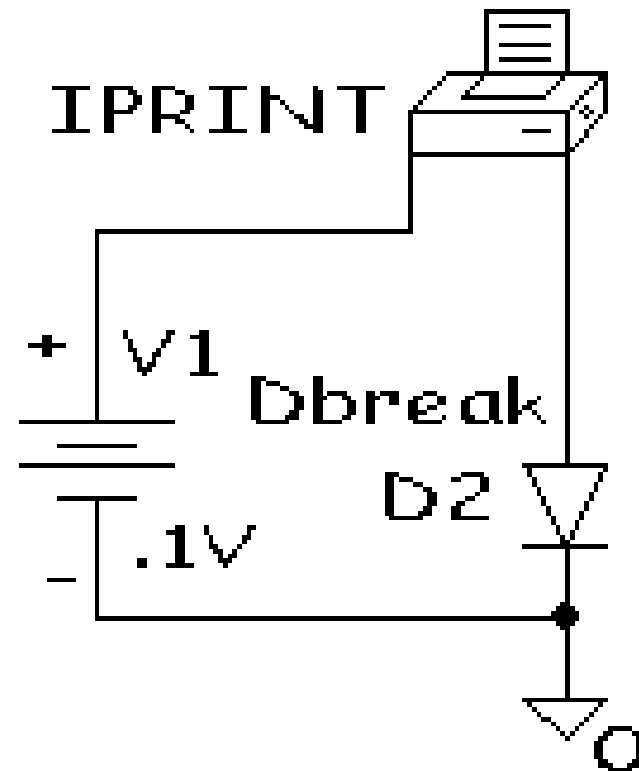
In the region where $i_D * R_S \gg V_d$.

$$di_D/V_d \sim 1/R_{S_{eff}}$$

$$dV_d/di_D \equiv R_{S_{eff}}$$

Getting Diode Data for Parameter Extraction

- The model used
.model Dbreak D(
Is=1e-13 N=1
Rs=.5 Ikf=5m
Isr=.11n Nr=2)
- Analysis has V1 swept, and IPRINT has V1 swept
- i_D , V_d data in Output



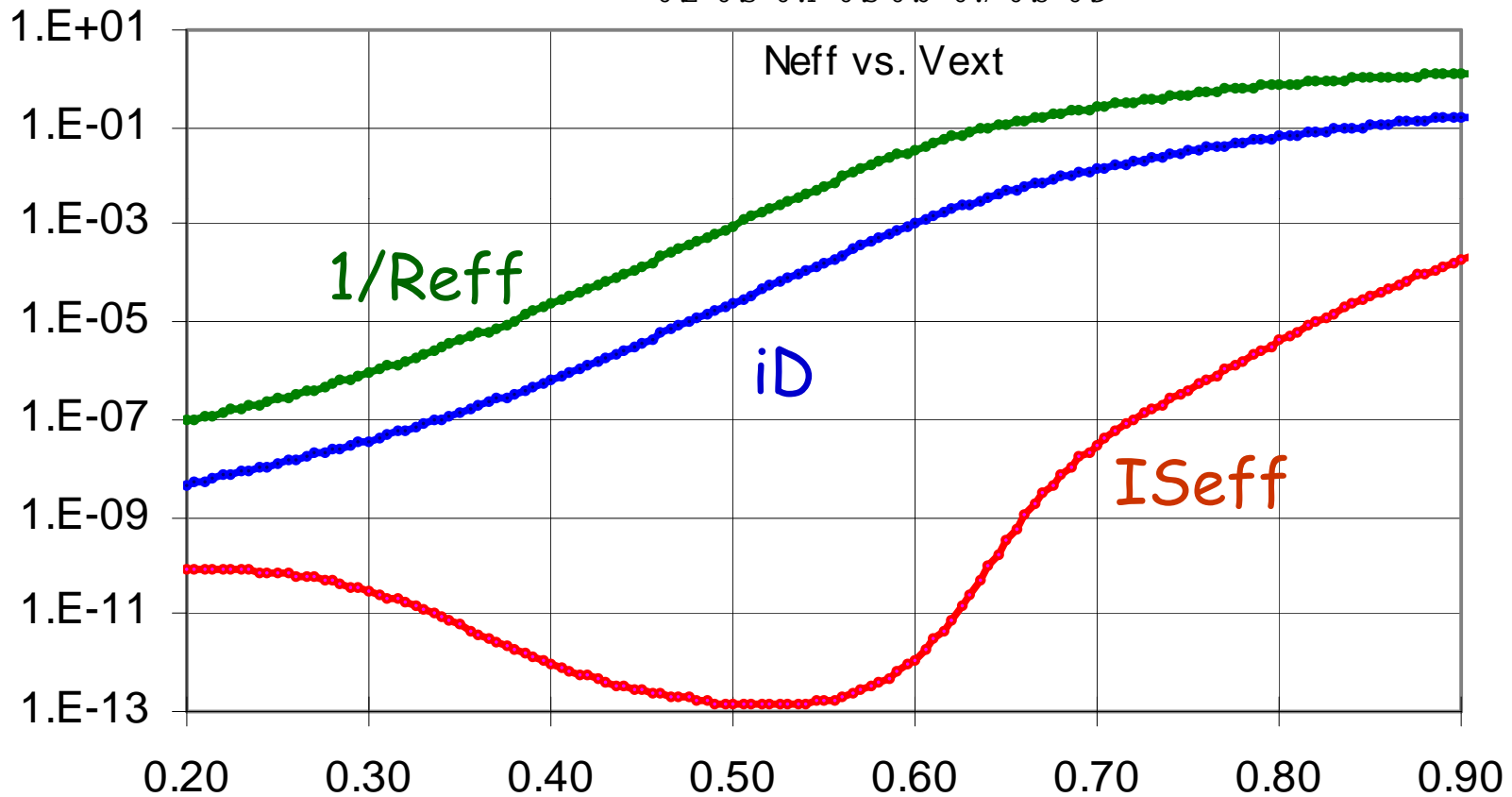
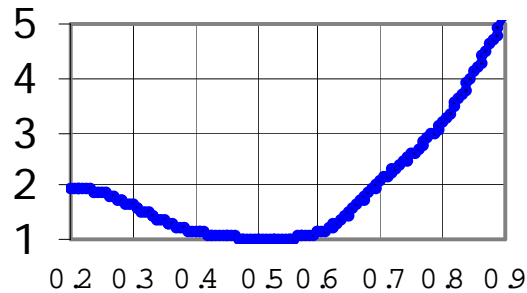
di_D/dV_d - Numerical Differentiation

V_d	i_D	di_D/dV_d (central difference)
$V_d(n-1)$	$i_D(n-1)$... etc. ...
$V_d(n)$	$i_D(n)$	$(i_D(n+1) - i_D(n-1))/(V_d(n+1) - V_d(n-1))$
$V_d(n+1)$	$i_D(n+1)$	$(i_D(n+2) - i_D(n))/(V_d(n+2) - V_d(n))$
$V_d(n+2)$	$i_D(n+2)$... etc. ...

$d/i_D/dV_d$ - Numerical Differentiation

V_d	i_D	$d/i_D/dV_d$ (central difference)
$V_d(n-1)$	$i_D(n-1)$... etc. ...
$V_d(n)$	$i_D(n)$	$/i_D(n+1)/i_D(n-1)/(V_d(n+1)-V_d(n-1))$
$V_d(n+1)$	$i_D(n+1)$	$/i_D(n+2)/i_D(n)/(V_d(n+2) - V_d(n))$
$V_d(n+2)$	$i_D(n+2)$... etc. ...

Diode Par. Extraction



L10 February 23 $iD(A)$, $Iseff(A)$, and $1/Reff(mho)$ vs. $Vext(V)$

Results of Parameter Extraction

- At $V_d = 0.2$ V, $N_{\text{Reff}} = 1.97$,
 $I_{\text{SReff}} = 8.99\text{E-}11$ A.
- At $V_d = 0.515$ V, $N_{\text{eff}} = 1.01$,
 $I_{\text{Seff}} = 1.35 \text{ E-}13$ A.
- At $V_d = 0.9$ V, $R_{\text{Seff}} = 0.725$ Ohm
- Compare to
 .model Dbreak D(Is=1e-13
 N=1 Rs=.5 I kf=5m I sr=.11n
 Nr=2)

Hints for RS and NF parameter extraction

In the region where $v_D > VKF$. Defining

$$v_D = v_{Dext} - i_D * RS \text{ and } I_{HLI} = [IS * IKF]^{1/2}.$$

$$i_D = I_{HLI} \exp(v_D / 2NV_+) + ISR \exp(v_D / NR V_+)$$

$$di_D / di_D = 1 \approx (i_D / 2NV_+) (dv_{Dext} / di_D - RS) + \dots$$

Thus, for $v_D > VKF$ (highest voltages only)

- plot i_D^{-1} vs. (dv_{Dext} / di_D) to get a line with
- slope = $(2NV_+)^{-1}$, intercept = $-RS / (2NV_+)$

Application of RS to lower current data

In the region where $v_D < VKF$. We still have

$$v_D = v_{Dext} - i_D * RS \text{ and since.}$$

$$i_D = IS \exp(v_D / NV_+) + ISR \exp(v_D / NRV_+)$$

- Try applying the derivatives for methods described to the variables i_D and v_D (using RS and v_{Dext}).
- You also might try comparing the N value from the regular N extraction procedure to the value from the previous slide.

References and Endnotes

- Where not otherwise noted, figures with a figure number (e.g., Fig 3.2) are taken from:
 - Electronics, 2nd edition, by Allan R. Hambley, Prentice Hall, Upper Saddle River, NJ, © 2000.
- (1) Microsim PSpice Reference Manual, Version 8.0, June, 1997. © 1997, MicroSim Corporation. [pspref.pdf]